

MECE 3350U  
Control Systems

Lecture 8  
Transient Response

## Videos in this lecture

Lecture: [https://youtu.be/J8jp\\_3KaXLw](https://youtu.be/J8jp_3KaXLw)

Exercise 35: <https://youtu.be/FgjQ0uxgwd0>

Exercise 36: <https://youtu.be/bhnWk-hTjgI>

Exercise 37: <https://youtu.be/a0Jt3LBX1Uw>

Exercise 38: <https://youtu.be/VdJisStqrx8>

Exercise 39: <https://youtu.be/1XTY1E604iE>

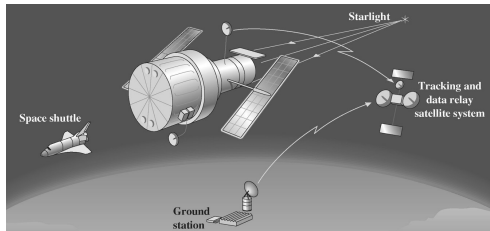
# Applications

The levitation control system of the train must ensure that the train does not touches the guide. How can we design a controller that reacts as fast as possible with no overshoot?

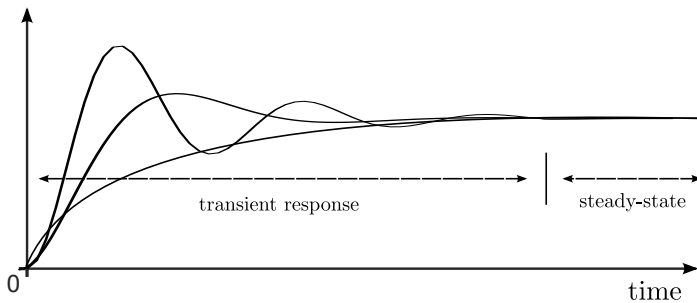


# Applications

The pointing control system of a space telescope is desired to achieve an accuracy of 0.01 minute of arc. How can we limit the steady state error while avoiding transient oscillations?

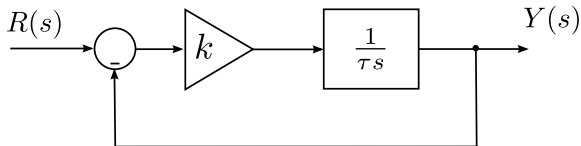


## Transient response



## First order systems

Consider the first order closed-loop system shown with a proportional gain  $k$



The transfer function  $Y(s)/R(s)$  is

$$\frac{Y(s)}{R(s)} = \frac{1}{\left(\frac{\tau}{k}\right)s + 1}$$

How does  $k$  influence the transient and steady state response?

To analyse the performance of the system, we need to specify a **standard test input signal**.

## Standard test signals

Impulse function

$$\delta(t) = \begin{cases} A & t = 0 \\ 0 & t \neq 0 \end{cases} \rightarrow I(s) = A$$

Step function

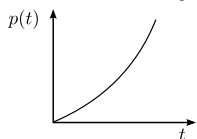
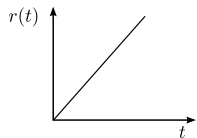
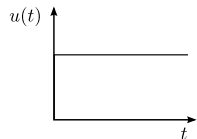
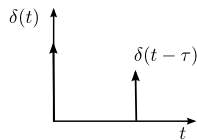
$$u(t) = \begin{cases} A & t \geq 0 \\ 0 & t < 0 \end{cases} \rightarrow U(s) = A \frac{1}{s}$$

Ramp function

$$r(t) = \begin{cases} At & t \geq 0 \\ 0 & t < 0 \end{cases} \rightarrow R(s) = A \frac{1}{s^2}$$

Parabolic function

$$p(t) = \begin{cases} A \frac{t^2}{2} & t \geq 0 \\ 0 & t < 0 \end{cases} \rightarrow P(s) = A \frac{1}{s^3}$$



## Temporal response

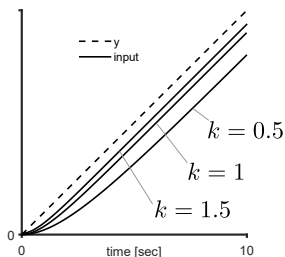
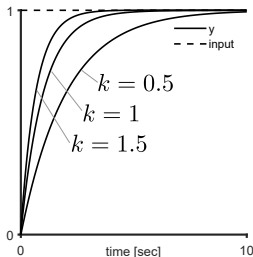
Step response  $r(t) = 1$

$$Y(s) = \frac{1}{s} \frac{1}{\left(\frac{\tau}{k}\right)s + 1} \rightarrow \mathcal{L}^{-1} \rightarrow y(t) = 1 - e^{-\frac{k}{\tau}t}$$

Ramp response  $r(t) = t$

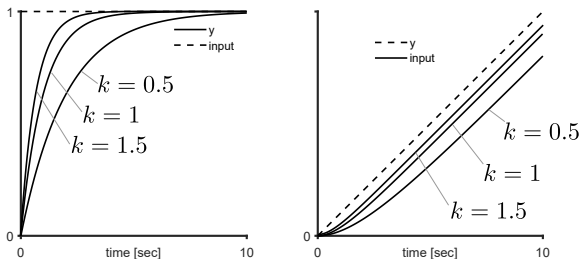
$$Y(s) = \frac{1}{s^2} \frac{1}{\left(\frac{\tau}{k}\right)s + 1} \rightarrow \mathcal{L}^{-1} \rightarrow y(t) = t - \frac{\tau}{k}(1 - e^{-\frac{k}{\tau}t})$$

Effects of  $k$  for  $k > 0$  for  $\tau = 1$ .





## Temporal response - first order system



In a first-order system:

- $k$  reduces the **time constant** of the system
- The higher  $k$ , the faster the response
- What is the maximum control-loop gain  $k$  ?

## Time constant - first order systems

Impulse:

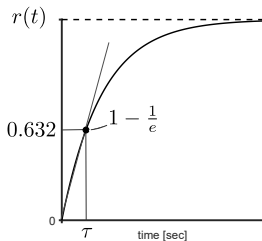
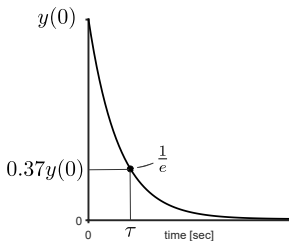
$$H(s) = \frac{1}{s\tau + 1} \rightarrow \mathcal{L}^{-1} \rightarrow y(t) = y(0)e^{-\frac{t}{\tau}}$$

→ When  $t = \tau$ , the response 37% ( $1/e$ ) of  $y(0)$

Step response

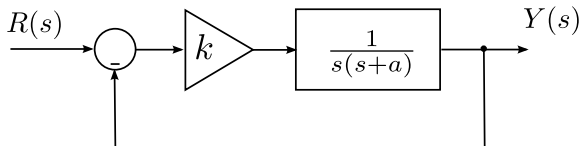
$$H(s) = \frac{1}{s} \frac{1}{s\tau + 1} \rightarrow \mathcal{L}^{-1} \rightarrow y(t) = 1 - e^{-\frac{t}{\tau}}$$

→ When  $t = \tau$ , the response 67% ( $1 - 1/e$ ) of its steady state value



## Second-order systems

Consider now the following second order control system:



The transfer function is

$$\frac{Y(s)}{R(s)} = \frac{k}{s^2 + sa + k}$$

We can rewrite the above equation in the standard formulation:

$$\frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

where:  $\omega_n = \sqrt{k}$ ,  $\zeta = a/(2\sqrt{k})$ .

## Transient response

$$\frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

where:  $\omega_n = \sqrt{k}$ ,  $\zeta = a/(2\sqrt{k})$ .

How does  $k$  influence the response of the system?

→ The natural frequency  $\omega_n$  depends on  $k$

→ The damping ratio  $\zeta$  depends on  $k$

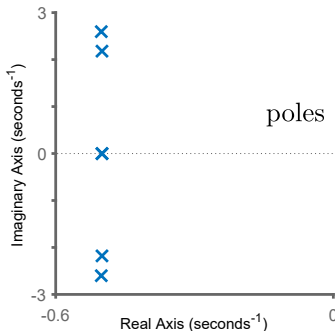
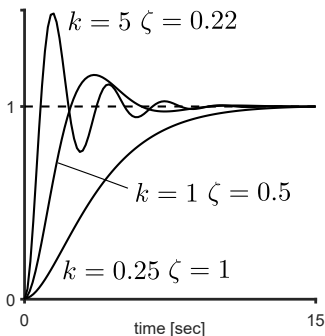
The response for an unit step input when  $0 < \zeta < 1$  is

$$y(t) = 1 - \frac{1}{\sqrt{1 - \zeta^2}} e^{-\zeta\omega_n t} \sin \left[ \left( \omega_n \sqrt{1 - \zeta^2} \right) t + \cos^{-1} \zeta \right]$$

## Transient response - second-order systems

$$\frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$y(t) = 1 - \frac{1}{\beta} e^{-\zeta\omega_n t} \sin\left(\omega_n \sqrt{1 - \zeta^2} t + \cos^{-1} \zeta\right) \quad (1)$$

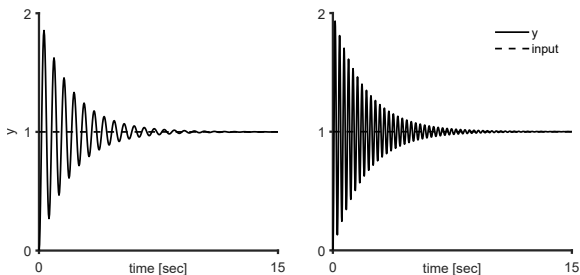


## Transient response - second-order systems

$$\frac{Y(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

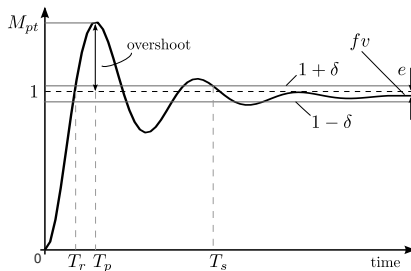
$$y(t) = 1 - \frac{1}{\sqrt{1-\zeta^2}} e^{-\zeta\omega_n t} \sin\left(\omega_n \sqrt{1-\zeta^2} t + \cos^{-1} \zeta\right)$$

Natural frequency for  $k = 1000$  and  $k = 5000$ .



How can we evaluate the performance of the controller?

## Measures of performance



- Rise time  $T_r$ , peak time  $T_p$ , and peak value  $M_{pt}$
- Settling time  $T_s$ :  $y(t)$  within 2% of its final value
- Percent overshoot  $P.O.$
- $T_r$  and  $T_p$  characterize the **swiftness** of the response
- $P.O.$  and  $T_s$  characterize the **closeness** of the response to the input

## Overshoot

For an unit step input, the percent overshoot is

$$P.O. = \frac{M_{pt} - f_v}{f_v} \times 100 \quad (2)$$

→  $M_{pt}$  is the peak value

→  $f_v$  is the magnitude of the input

Differentiating Eq (1) and setting it to zero yields the peak time as

$$T_p = \frac{\pi}{\omega_n \sqrt{1 - \zeta^2}} \quad (3)$$

Replacing (3) into (1) gives the peak response:

$$M_{pt} = 1 + e^{-\zeta \pi / \sqrt{1 - \zeta^2}} \quad (4)$$

Thus, the percentage overshoot ( $f_v = r(t) = 1$ ) is

$$P.O. = 100e^{-\zeta \pi / \sqrt{1 - \zeta^2}} \quad (5)$$



## Settling time

For an unit step input and  $0 < \zeta < 1$ , recall that

$$y(t) = 1 - \frac{1}{\sqrt{1 - \zeta^2}} e^{-\zeta \omega_n t} \sin \left( \omega_n \sqrt{1 - \zeta^2} t + \cos^{-1} \zeta \right)$$

When  $t = T_s$ , the response is within 2% of its final value, thus:

$$e^{-\zeta \omega_n T_s} < 0.02$$

or

$$\zeta \omega_n T_s \approx 4 \quad (6)$$

therefore

$$T_s = \frac{4}{\zeta \omega_n} = 4\tau \quad (7)$$

where  $\tau = 1/\zeta \omega_n$  is the time constant.

The settling time is equal to 4 times the time constant

# Settling time

The characteristic equation

$$s^2 + 2\zeta\omega_n s + \omega_n^2$$

has poles:

$$s_1 = -\zeta\omega_n + j\omega_n\sqrt{1 - \zeta^2}$$

$$s_1 = -\zeta\omega_n - j\omega_n\sqrt{1 - \zeta^2}$$

or  $s = -\zeta\omega_n \pm \omega_n\sqrt{\zeta^2 - 1}$ . Since

$$T_s \approx \frac{4}{\zeta\omega_n} \quad (8)$$

Therefore the settling time is inversely proportional to the real part of the poles.

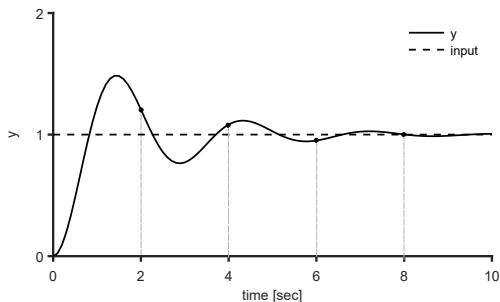
## Settling time

Consider the function

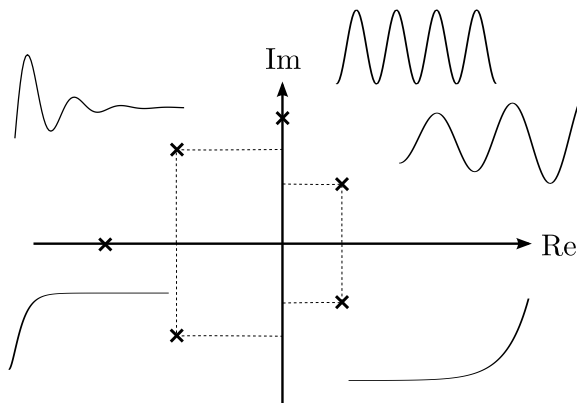
$$H = \frac{5}{s^2 + s + 5}$$

thus:  $\omega_n = \sqrt{5}$  and  $\zeta = 1/(2\sqrt{5})$ . The time constant is

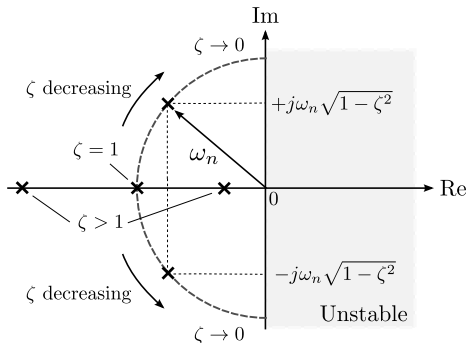
$$\tau = \frac{1}{\zeta\omega_n} = 2 \text{ sec}$$



## Poles and transient response



Damping ratio	Roots	Systems response
$\zeta > 1$	Distinct real	overdamped
$\zeta = 1$	Equal real	damped
$0 < \zeta < 1$	Complex conjugate	underdamped
$\zeta = 0$	Purely imaginary	undamped
$\zeta < 0$	Positive	unstable



## Exercise 35

A feedback system with a negative unity feedback has the loop transfer function

$$L(s) = C(s)G(s) = \frac{2(s+8)}{s(s+4)}.$$

Determine:

- **(a)** The closed-loop transfer function
- **(b)** The time response for an input  $r(t) = A$
- **(c)** The percent overshoot of the response
- **(d)** The steady state error

## Exercise 35 - continued

(a) The closed-loop transfer function

$$L(s) = C(s)G(s) = \frac{2(s+8)}{s(s+4)}.$$

## Exercise 35 - continued

(b) The time response

$$Y(s) = \frac{A}{s} \frac{2(s+8)}{s^2 + 6s + 16}$$



## Exercise 35 - continued

(c) The percentage overshoot

$$Y(s) = \frac{A}{s} \frac{2(s+8)}{s^2 + 6s + 16}$$

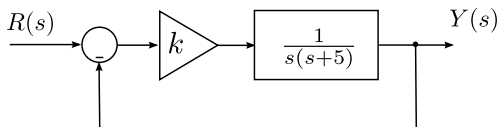
## Exercise 35 - continued

**(d)** The steady state error

$$Y(s) = \frac{A}{s} \frac{2(s+8)}{s^2 + 6s + 16}$$

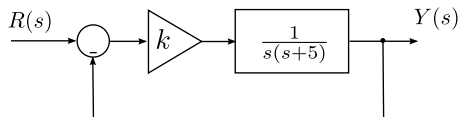
## Exercise 36

Consider the following block diagram:



- **(a)** Calculate the steady-state error for a ramp input
- **(b)** Select  $k$  that will result in zero overshoot to a step input

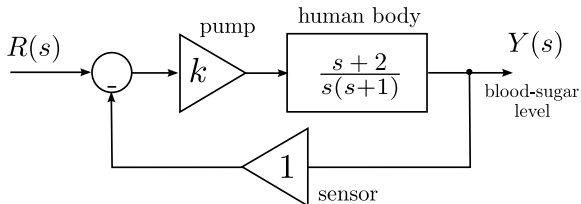
## Exercise 36 - continued



## Exercise 36 - continued

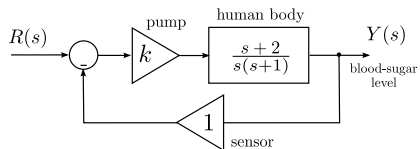
## Exercise 37

An insulin pump injection system for diabetic persons has a feedback control as shown.

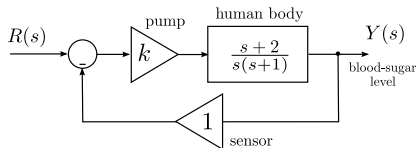


Calculate a suitable gain  $k$  so that the percent overshoot of the step response due to the drug injection is 7%.  $R(s)$  is the desired blood sugar level and  $Y(s)$  is the actual level. Plot the expected overshoot for different  $k$  using Matlab.

## Exercise 37 - continued



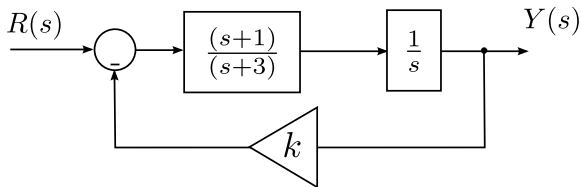
## Exercise 37 - continued





## Exercise 38

Consider the following closed loop system

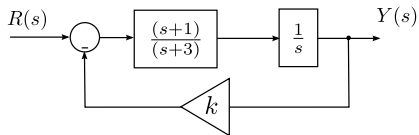


Determine:

- **(a)** Determine the closed loop transfer function
- **(b)** Select  $k$  so that the steady state error to a unit step input is bounded.

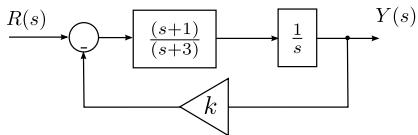
## Exercise 38 - continued

(a) Determine the closed loop transfer function



## Exercise 38 - continued

**(b)** Determine the steady state error to a unit ramp input

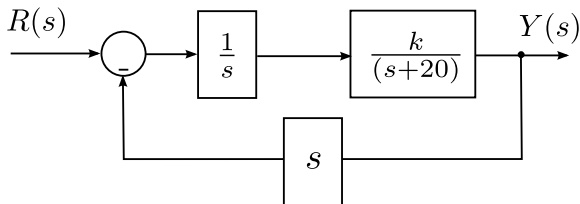


## Exercise 38 - continued

(c) Determine  $k$  so that  $e = 0$  when  $r(t) = 1$

## Exercise 39

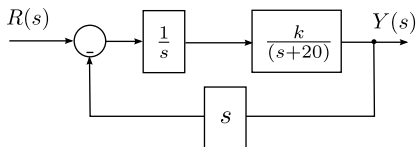
A closed-loop system designed to orient a photovoltaic array towards the direction of maximum solar incidence has the following structure:



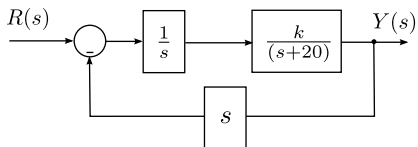
If  $k = 20$ , determine:

- **(a)** The time constant of the closed loop system
- **(b)** The settling time to within 2% of the final value of the system to an unit step **disturbance**.

## Exercise 39 - continued



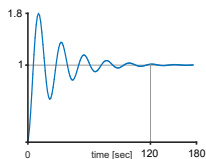
## Exercise 39 - continued



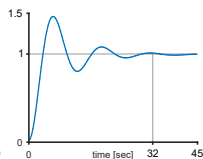
## Skills check 22 - From 2018 midterm examination

Which of the following time domain signals best describes the response of the function  $H(s)$  given below to a unit step-type input? (4 marks/100).

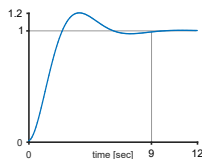
$$H(s) = \frac{1}{4s^2 + s + 1}$$



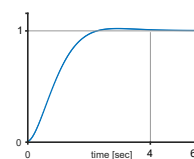
(a)



(b)



(c)



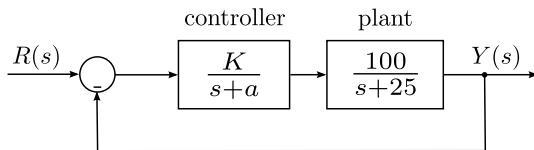
(d)

Answer in the last slides



## Skills check 23 - From 2018 final examination

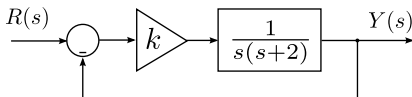
A controller is to be designed for a plant as shown.



- (a) Find the transfer function  $Y(s)/R(s)$ .
- (b) Specify the controller gain  $K$  and the value of  $a$  so that the overall closed-loop response to a unit step input has an overshoot of no more than 25% and a settling time of no more than 0.1 seconds.

## Skills check 24 - From 2018 midterm examination

A robotic system has been designed to operate on a beating heart by ensuring continuous contact between a surgical tool and the heart. The block diagram of such a system is shown below where  $R(s)$  is the desired position of the surgical tool i.e., the position of the epicardium (heart wall), and  $Y(s)$  is the actual tool position. Any controller overshoot would result in the tool puncturing the heart.



- (a) Determine the steady-state error for an unit step input of  $R(s)$  as a function of  $k$ .
- (b) If  $k = 10$ , what is the expected overshoot ?
- (c) Determine the maximum value of  $k$  that results in no overshoot.
- (d) What is the settling time as a function of  $k$  ?

## Answers to skills check

SC 22 - (b)

SC 23 - In (a) the transfer function is

$$\frac{Y(s)}{R(s)} = \frac{100k}{s^2 + s(a + 25) + 25a + 100k}$$

For question (b), we get  $k \approx 85$  and  $a \approx 55$ .

- SC 24 - (a)  $e_s s = 0 \forall k$ ,  
(b) P.O. = 35%,  
(c)  $k < 1$ ,  
(d)  $T_s = 4 \text{ sec } \forall k$

## Next class...

- Dominant poles