

MECE 3350U
Control Systems

Lecture 4
Transfer Functions

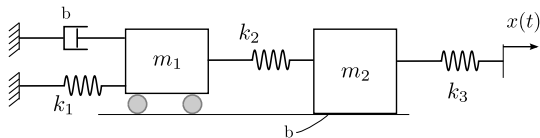
Outline of Lecture 4

By the end of today's lecture you should be able to

- Understand the concept of transfer functions
- Find the transfer function of a given system
- Find the temporal response of a system for a given input

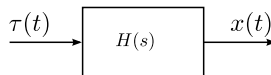
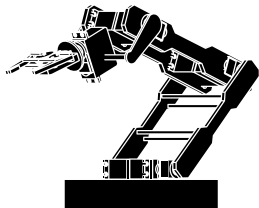
Applications

If the spring is stretched to a point $x(t) = 5$ mm, held, then released at time $t = 0$, how does the position of m_1 evolve in time?



Applications

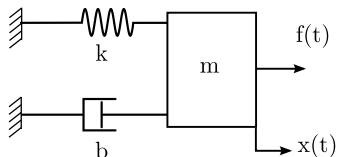
What torque must be applied to each of the robot joints so that end-effector moves along a given trajectory with a given speed?



Input/output relation

Transfer function: A relation between the input and output of a linear system

Example: Input: force $f(t)$, output: displacement $x(t)$



$$m \frac{d^2x}{dt^2} + b \frac{dx}{dt} + kx = f(t)$$

Taking the Laplace transform:

$$ms^2 X(s) + b s X(s) + k X(s) = F(s)$$

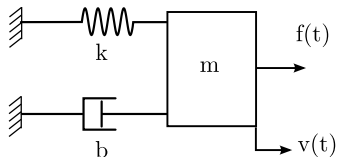
The transfer function is

$$X(s) [ms^2 + bs + k] = F(s)$$

$$H(s) = \frac{X(s)}{F(s)} = \frac{1}{ms^2 + bs + k} \quad // \quad (1)$$

Input/output relation

Example: Input: force $f(t)$, output: **velocity** $v(t)$



The dynamic model is $m \frac{dv}{dt} = f(t) - k \int v dt - b v$ (2)

Laplace transform of (2) is $m s V(s) + b V(s) + K \cdot \frac{1}{s} V(s) = F(s)$

The transfer function is

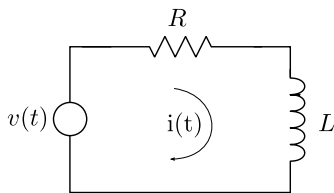
$$H(s) = \frac{V(s)}{F(s)} = \frac{s}{m s^2 + b s + k}$$

Handwritten note: A red circle around the s in the numerator is labeled "derivative of Eq (1)!" with an arrow pointing to it.

(3)

Input/output relation

Input: Voltage $v(t)$, output: Current $i(t)$



$$v(t) = Ri(t) + L \frac{di(t)}{dt} \quad (4)$$

Taking the Laplace transform of (4):

$$V(s) = RI(s) + LsI(s)$$

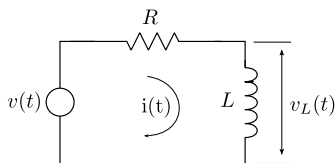
$$V(s) = I(s)[R + Ls]$$

The transfer function is

$$H(s) = \frac{I(s)}{V(s)} = \frac{1}{Ls + R}$$

Input/output relation

Input: Voltage $v(t)$, **output:** $v_L(t)$



$$v(t) = Ri(t) + L \frac{di(t)}{dt}, \quad v_L = L \frac{di(t)}{dt} \quad (5)$$

Taking the Laplace transform of (5):

$$V(s) = [R + Ls] I(s)$$

$$V_L(s) = Ls I(s)$$

$$I(s) = \frac{V_L(s)}{Ls}$$

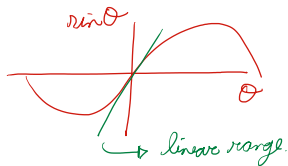
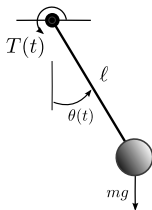
The transfer function is

$$\frac{I(s)}{V(s)} = \frac{1}{Ls + R}$$

$$\frac{V_L(s)}{V(s)} = \frac{Ls}{Ls + R} \rightarrow I(s)$$

Input/output relation

The equation of motion of the simple pendulum is



$$m \frac{d^2\theta(t)}{dt^2} + m \frac{g}{l} \sin \theta = T(t) \quad (6)$$

For small angles, $\sin \theta \approx \theta$, in the frequency domain:

$$ms^2 \theta(s) + \frac{mg}{l} \theta(s) = T(s)$$

$$\frac{\theta(s)}{T(s)} = \frac{1}{ms^2 + \frac{mg}{l}} \quad (7)$$

Transfer function poles and zeros

Transfer function: A rational function in the complex variable $s = \sigma + j\omega$:

$$H(s) = \frac{N(s)}{D(s)} = \frac{b_m s^m + b_{m-1} s^{m-1} + \dots + b_1 s + b_0}{a_n s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0} \quad (8)$$

The zeros z_i are the roots of

$$N(s) = 0$$

Thus:

$$\lim_{s \rightarrow z_i} N(s) = 0 \quad (9)$$

The poles p_i are the roots of

$$D(s) = 0$$

Thus:

$$\lim_{s \rightarrow p_i} H(s) = \infty \quad (10)$$

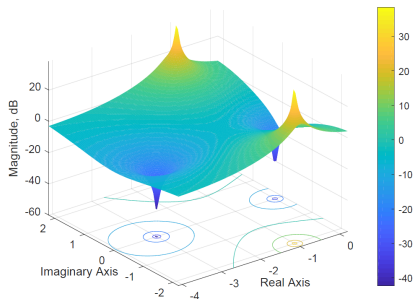
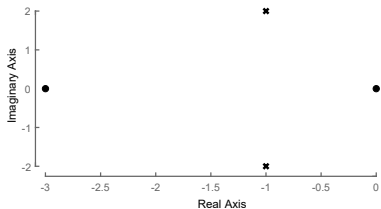
Poles and zeros

Consider the following function:

$$F(s) = \frac{s(s+3)}{s^2 + 2s + 5}$$

→ Poles: $-1 + 2j$, $-1 - 2j$

→ Zeros: 0 , -3



First order transfer functions

Order: The number of the highest derivative in the denominator (power of s)

Standard form of a first order system:

$$G(s) = k \frac{1}{\tau s + 1}$$

Characteristic equation: The denominator of the transfer function

$$H(s) = \frac{I(s)}{V(s)} = \frac{1}{Ls + R} = \frac{1}{R} \frac{1}{\frac{L}{R}s + 1}$$

Time constant: Characterizes the response to a step input of a first-order system.

$$\tau = \frac{L}{R} \tag{11}$$

→ The denominator must be in the form of $\tau s + 1$

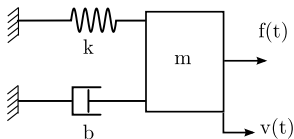
Second order transfer functions

Standard form

$$H(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \quad (12)$$

Where: ω_n is the natural frequency, ζ is the damping ratio.

We will come back to these definitions in the next lecture.



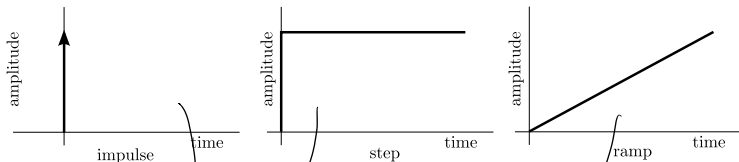
$$H(s) = \frac{X(s)}{F(s)} = \frac{1}{ms^2 + bs + k} = \frac{1}{m} \frac{1}{\left(s^2 + \frac{b}{m}s + \frac{k}{m}\right)} \Rightarrow \frac{1}{m} \cdot \frac{m}{k} \frac{\frac{k}{m}}{s^2 + \frac{b}{m}s + \frac{k}{m}}$$

$\Rightarrow \zeta = \frac{b}{2\sqrt{mk}}$: Dimensionless **damping ratio**

$\Rightarrow \omega_n = \sqrt{\frac{k}{m}}$: Natural frequency (rad/s)

Temporal response

Step 1: Define the input signal in Laplace domain



input signal time domain frequency domain

impulse $r(t) = \delta(t)$ $R(s) = 1$

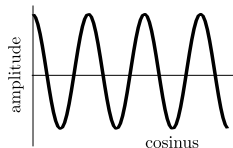
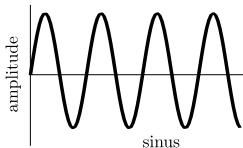
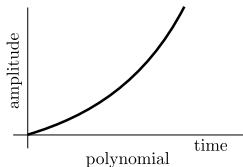
step $r(t) = a$ $R(s) = a \frac{1}{s}$

ramp $r(t) = at$ $R(s) = a \frac{1}{s^2}$

a is a constant

Temporal response

Step 1: Define the input signal in Laplace domain




input signal time domain frequency domain

polynomial $r(t) = at^n$ $R(s) = a \frac{n!}{s^{n+1}}$

sine $r(t) = \sin(at)$ $R(s) = \frac{a}{s^2+a^2}$

cosine $r(t) = \cos(at)$ $R(s) = \frac{s}{s^2+a^2}$

a is a constant

 \uparrow there is a typo in the notes.

Temporal response

Step 2: Replace the input signal in the transfer function

$$H(s) = \frac{X(s)}{F(s)} = \frac{1}{ms^2 + bs + k} \quad (13)$$

For a impulse input $f(t) = \delta(t)$, $F(s) = 1$ and the temporal response is

$$X(s) = \frac{1}{ms^2 + bs + k} \quad (14)$$

For a step-type input $f(t) = 1 \text{ N}$, $F(s) = 1/s$ and the temporal response is

$$X(s) = \frac{1}{ms^2 + bs + k} \left(\frac{1}{s} \right) \quad (15)$$

For a sinusoidal input $f(t) = 5 \sin(t) \text{ N}$:

$$X(s) = \frac{1}{ms^2 + bs + k} \left(\frac{5}{s^2 + 1} \right) \quad (16)$$

Temporal response

Step 3: Calculate the inverse transform of the resulting function

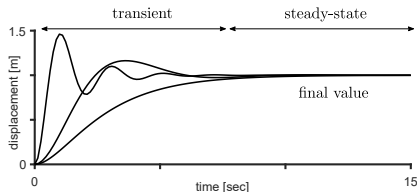
For a impulse input $f(t) = \delta(t)$, $F(s) = 1$ and the temporal response is

$$x(t) = \mathcal{L}^{-1} \left\{ \frac{1}{ms^2 + bs + k} \right\} \quad (17)$$

For a step-type input $f(t) = 1 \text{ N}$, $F(s) = 1/s$ and the temporal response is

$$x(t) = \mathcal{L}^{-1} \left\{ \frac{1}{ms^2 + bs + k} \left(\frac{1}{s} \right) \right\} \quad (18)$$

and so on.



Steady-state value

Final value theorem: Gives the steady-state value without computing the inverse transform.

If the function converges, i.e., the poles of $sX(s)$ have negative real parts, then:

$$\lim_{t \rightarrow \infty} x(t) = \lim_{s \rightarrow 0} sX(s) \quad (19)$$

For a step type input, the mass spring damper system settles at

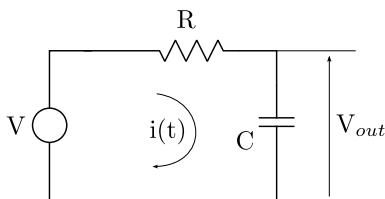
$$\lim_{t \rightarrow \infty} x(t) = \lim_{s \rightarrow 0} sX(s) = \lim_{s \rightarrow 0} s \left\{ \frac{1}{ms^2 + bs + k} \left(\frac{1}{s} \right) \right\} = \frac{1}{k} \quad (20)$$

For a impulse input, the *RL* system converges at

$$\lim_{t \rightarrow \infty} i(t) = \lim_{s \rightarrow 0} sI(s) = \lim_{s \rightarrow 0} s \left\{ \frac{1}{Ls + R} \right\} = 0 \quad (21)$$

Exercise 16

Find the transfer function $H(s)$ between the input voltage V and the output voltage V_{out} .



Procedure:

- Find the differential equation for the current
- Find the equation for the output voltage
- Calculate the transfer function

Exercise 16 - continued

$$v(t) = i(t)R + \frac{1}{C} \int i(t) dt$$

$$V(s) = \left[R + \frac{1}{Cs} \right] I(s) \quad (1)$$

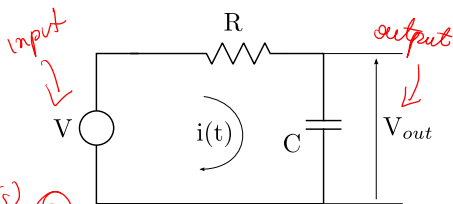
$$V_{out}(t) = \frac{1}{C} \int i dt \rightarrow V_{out}(s) = \frac{1}{Cs} I(s) \quad (2)$$

$$(2) \rightarrow I(s) = V_{out}(s) C s \rightarrow (1)$$

$$V(s) = \left[R + \frac{1}{Cs} \right] V_{out}(s) C s$$

$$V(s) = (RCs + 1) V_{out}(s)$$

$$\frac{V_{out}(s)}{V(s)} = \frac{1}{RCs + 1}$$



final value.

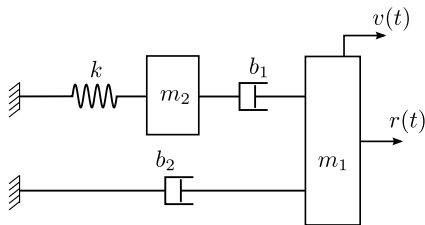
for an impulse $\rightarrow \begin{cases} v(t) = \delta(t) \\ V(s) = 1 \end{cases}$

$$v_{out}(t) = V_{out}(s) \times s \Big|_{s \rightarrow 0}$$

$$v_{out}(t) = \left(\frac{1}{RCs + 1} \times 1 \right) \times s \Big|_{s \rightarrow 0} = 0 //$$

Exercise 17

Find the transfer function $H(s) = \frac{V(s)}{R(s)}$ between the force $r(t)$ and the velocity of mass m_1 .



Procedure:

- Find the differential equation the velocity of each mass
- Calculate the Laplace transform
- Calculate the transfer function

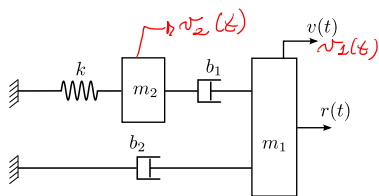
Exercise 17 - continued

Mass 1

$$r(t) - b_2 v_1 - b_1 (v_1 - v_2) = m_1 \dot{v}_1$$

$$R(s) - b_2 V_1(s) - b_1 (V_1(s) - V_2(s)) = m_1 s V_1(s)$$

$$\boxed{[m_1 s + b_1 + b_2] V_1(s) - b_1 V_2(s) = R(s) \quad (1)}$$



Mass 2

$$b_1 (v_1 - v_2) - b_1 \int v_2 dt = m_2 \frac{dv_2}{dt}$$

$$m_2 s V_2(s) + b_1 (V_2(s) - V_1(s)) + \frac{k}{s} V_2(s) = 0$$

$$\boxed{-b_1 V_1(s) + \left(m_2 s + b_1 + \frac{k}{s}\right) V_2(s) = 0 \quad (2)}$$

Exercise 17 - continued

$$\textcircled{1} [m_1 s + (b_1 + b_2)] V_1(s) - b_1 V_2(s) = R(s)$$

$$\textcircled{2} -b_1 V_1(s) + \left(m_2 s + b_1 + \frac{k}{s} \right) V_2(s) = 0$$

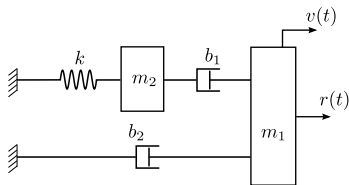
$$\hookrightarrow V_2(s) = \frac{b_1}{m_2 s + b_1 + \frac{k}{s}} V_1(s) \quad \textcircled{3}$$

$\textcircled{3}$ in $\textcircled{1}$ gives

$$[m_1 s + (b_1 + b_2)] V_1(s) - b_1 \left(\frac{b_1}{m_2 s + b_1 + \frac{k}{s}} \right) V_1(s) = R(s)$$

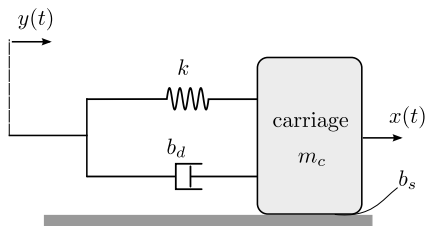
↙ solving.

$$\frac{V_1(s)}{R(s)} = \frac{m_2 s^2 + b_1 s + k}{(m_1 s + b_1 + b_2)(m_2 s^2 + b_1 s + k) - b_1^2 s}$$



Exercise 18

A high precision positioning slide is shown in the figure. The drive shaft friction is $b_d = 0.65$, the drive shaft spring constant is $k = 1.8$, $m_c = 1$, and the slide friction is $b_s = 0.9$.



Determine:

- Find the transfer function $H(s) = X(s)/Y(s)$.
- Calculate the natural frequency, damping ratio, the poles, and zeros of $H(s)$
- Find the steady-state value for a step input
- Plot the step response of using Matlab

Exercise 18 - continued

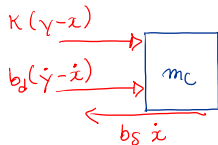
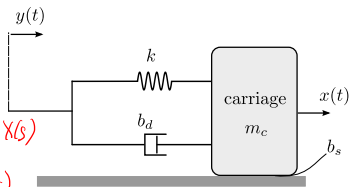
$$m \ddot{x} = k(y - x) + b_d(\dot{y} - \dot{x}) - b_s \dot{x}$$

$$m s^2 X(s) = k Y(s) - k X(s) + b_d s Y(s) - b_d s X(s) - b_s X(s)$$

$$m s^2 X(s) + (b_s + b_d) s X(s) + k X(s) = b_d s Y(s) + k Y(s)$$

$$X(s) [m s^2 + (b_s + b_d) s + k] = Y(s) [b_d s + k]$$

$$\frac{Y(s)}{X(s)} = \frac{b_d s + k}{m s^2 + (b_s + b_d) s + k}$$



Exercise 18 - continued

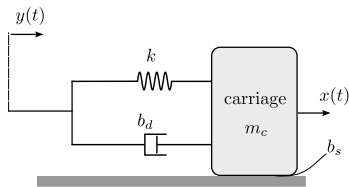
$$\frac{Y(s)}{X(s)} = \frac{b_d s + K}{m s^2 + (b_s + b_d) s + K}$$

for a step input $X(s) = \frac{1}{s}$

$$Y(s) = \frac{b_d s + K}{m s^2 + (b_s + b_d) s + K} \times \frac{1}{s}$$

final value $\Rightarrow s Y(s) \Big|_{s \rightarrow 0}$

$$x(t) = \lim_{t \rightarrow \infty} s Y(s) \Big|_{s \rightarrow 0} = \cancel{s} \cdot \frac{b_d s + K}{m s^2 + (b_s + b_d) s + K} \cdot \frac{1}{\cancel{s}} \Big|_{s \rightarrow 0} = \frac{K}{K} = 1 //$$



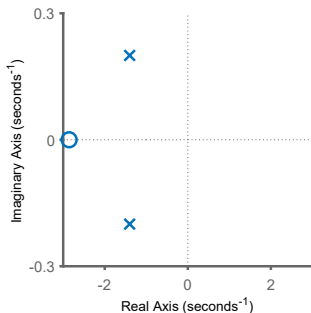
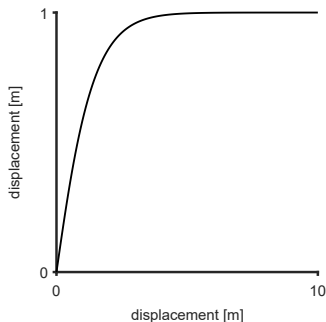
Exercise 18 - continued - Using Matlab

$H = \text{tf}([0.7 \ 2],[1 \ 2.8 \ 2])$ → Transfer function
replace with numbers given in the problem.

$\text{damp}(H)$ → Natural frequency and damping

$\text{step}(H,10)$ → Step response

$\text{pzplot}(H)$ → Location of zeros and poles



Exercise 19

Calculate the natural frequency and damping ratio of the following transfer function

$$T(s) = \frac{1.05 \times 10^7}{2s^2 + 1.6 \times 10^3s + 1.05 \times 10^7}$$

Determine:

- Write the transfer function in standard form
- Find the steady-state value for a step input
- Calculate the natural frequency and damping ratio

Exercise 19 - continued

$$T(s) = \frac{1.05 \times 10^7}{2s^2 + 1.6 \times 10^3 s + 1.05 \times 10^7}$$

$$T(s) = \frac{0.525 \times 10^7}{s^2 + 0.8 \times 10^3 s + 0.525 \times 10^7}$$

$$\omega_n^2 = 0.525 \times 10^7, \quad \boxed{\omega_n = 2291 \text{ rad/s}}$$

$$2\zeta\omega_n = 0.8 \times 10^3$$

$$\zeta = \frac{0.8 \times 10^3}{2 \times 2291} \rightarrow \boxed{\zeta = 0.17}$$

for on step input

$$sT(s) \Big|_{s \rightarrow 0}$$

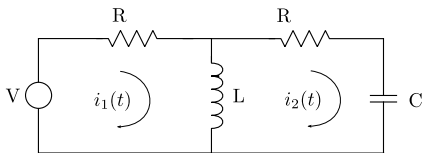
$$s \frac{1.05 \times 10^7}{2s^2 + 1.6 \times 10^3 s + 1.05 \times 10^7} \cdot \frac{1}{s} \Big|_{s \rightarrow 0}$$

$$= 1 //$$

Exercise 20

Find the transfer function $G(s) = \frac{I_2(s)}{V(s)}$ of the circuit shown. Then, calculate the step response of the circuit using Matlab. Take $R = 10 \Omega$, $C = 0.001 \text{ F}$, $L = 0.1 \text{ H}$, $V = 5 \text{ V}$.

*Homework
Try this on your own.*



Determine:

- Find the transfer function $H(s) = I_2(s)/V(s)$.
- Find the steady-state value for a step input
- Plot the step response of using Matlab

Exercise 20 - continued

$$V(s) = I_1(s)R + Ls(I_1(s) - I_2(s)) \quad (1)$$

$$I_2(s) = \frac{V + I_2 L}{R + sL}$$

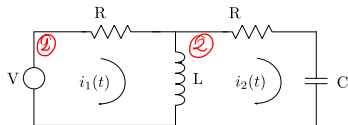
replace

$$[I_2(s) - I_1(s)]sL + I_2(s)R + \frac{1}{Cs}I_2(s) = 0$$

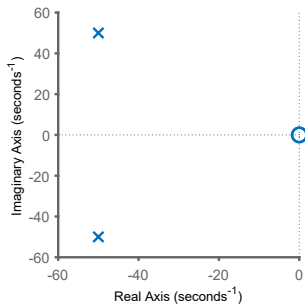
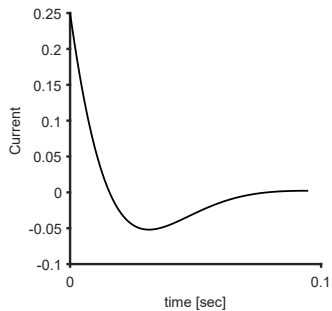
$$(I_2(s)) \left[sL + R + \frac{1}{Cs} \right] - sL \left(\frac{V + I_2 L}{R + sL} \right) = 0$$

$$R_2 + \frac{1}{Cs} + Ls - \frac{L^2 s^2}{R + sL} = \frac{Ls}{R + sL} V(s) \cdot \frac{1}{I_2}$$

$$\frac{I_2(s)}{V(s)} = \frac{LCs^2}{s^2(2RCL) + s(L + R^2C) + R}$$



Exercise 20 - continued - Using Matlab



Next class...

- Effect of pole locations